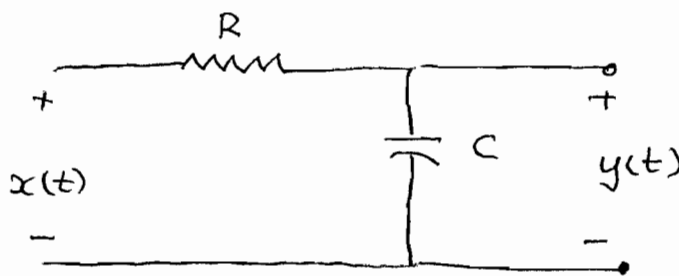


ECE-314, FALL 2008

SIGNALS & SYSTEMS

On the ideal LPF:

In class we have looked at the R-C voltage divider circuit as a LPF:



The corresponding frequency response:

$$H(j\omega) = \frac{a}{a + j\omega}, \quad \omega \in \mathbb{R}, \quad a = \frac{1}{RC}$$

$$|H(j\omega)| = \frac{a}{\sqrt{a^2 + \omega^2}}, \quad \omega_{3dB} = \pm a$$

$$\arg(H(j\omega)) = -\tan^{-1}\left(\frac{\omega}{a}\right)$$

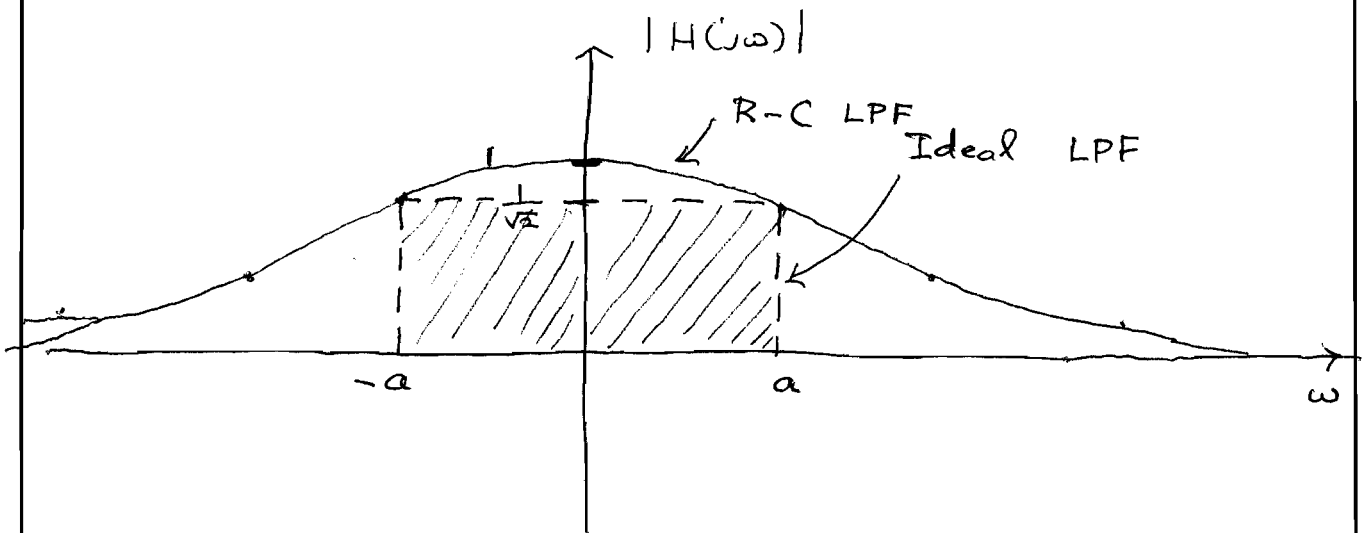
However, this is not the response of the ideal LPF that rejects signals completely outside of  $\omega \in [-a, a]$

3-0235 — 50 SHEETS — 5 SQUARES  
3-0236 — 100 SHEETS — 5 SQUARES  
3-0237 — 200 SHEETS — 5 SQUARES  
3-0137 — 200 SHEETS — FILLER

COMET

Suppose we consider a LPF with frequency response:

$$H_{LP}(j\omega) = \begin{cases} 1, & |\omega| < a \\ 0, & \text{otherwise} \end{cases}$$



The corresponding impulse responses:

$$h_{RC}(t) = \frac{1}{RC} e^{-t/RC} u(t)$$

$$h_{LP}(t) = \frac{1}{2\pi} \int_{-a}^a e^{j\omega t} d\omega = \frac{1}{2\pi} \frac{e^{j\omega t}}{jt} \Big|_{-a}^a$$

$$h_{LP}(t) = \frac{a}{\pi} \frac{\sin(at)}{at}, \quad t \in \mathbb{R}$$

Note first the sharp response of the ideal LPF versus the slow decay of the R-C lowpass filter.

3-0235 — 50 SHEETS — 5 SQUARES  
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 3-0137 — 200 SHEETS — FILLER

COMET

- The second but more serious implication is in regards to causality:

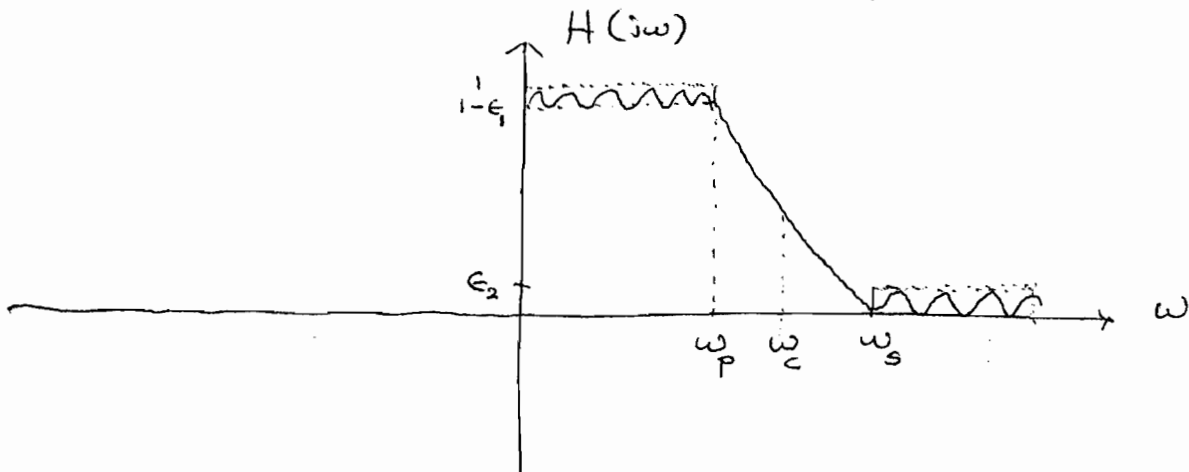
$$h_{RC}(t) = \frac{1}{RC} e^{-t/RC} u(t) = 0, \quad t < 0$$

$$h_{LP}(t) \neq 0, \quad t < 0$$

The ideal LPF is non-causal and therefore not implementable.

- The third obvious difference is in the time-domain; where  $h_{RC}(t)$  decays to zero faster than  $h_{LP}(t)$  because of duality

Practical LPF response:



- $\epsilon_1$ : Passband tolerance
- $\epsilon_2$ : Stopband tolerance
- $\omega_p$ : Passband freq
- $\omega_s$ : Stopband freq.