

Modulation of Random Processes

Let us now consider the modulation of a random process $X(t)$ with the random-phase sinusoidal signal described before :

$$Y(t) = X(t) \cos(\Omega_c t + \Theta), \quad (51)$$

where $\Theta \sim \mathbf{U}([0, 2\pi])$ and $X(t)$ and Θ are independent random variables $\forall t \in \mathbf{R}$. The ACF of the process $Y(t)$ is given by :

$$R_{YY}(t_1, t_2) = E[Y(t_1)Y^*(t_2)] = E[X(t_1) \cos(\Omega_c t_1 + \Theta) X^*(t_2) \cos(\Omega_c t_2 + \Theta)] \quad (52)$$

Using the independence of $X(t)$ & Θ we have :

$$\begin{aligned} R_{YY}(t_1, t_2) &= E_{\Theta} [X(t_1)X^*(t_2)] E_{\Theta} [\cos(\Omega_c(t_1 + t_2) + 2\Theta) \cos[\Omega_c(t_1 - t_2)]] \\ R_{YY}(t_1, t_2) &= \left(\frac{R_{XX}(t_1, t_2)}{2} \right) \{E_{\Theta} \{\cos[\Omega_c(t_1 + t_2) + 2\Theta]\} + E_{\Theta} \{\cos[\Omega_c(t_1 - t_2)]\}\} \end{aligned} \quad (53)$$

Since $\Theta \sim \mathbf{U}([0, 2\pi])$ the first expectation becomes zero and then we have :

$$R_{YY}(t_1, t_2) = \frac{R_{XX}(t_1, t_2)}{2} \cos[\Omega_c(t_1 - t_2)] \quad (54)$$

This relation implies that if $X(t)$ is a WSS random process then $Y(t)$ is also a WSS process, hence :

$$R_{YY}(\tau) = \frac{R_{XX}(\tau)}{2} \cos(\Omega_c \tau) \quad (55)$$

Modulation of the $X(t)$ with the random cosine results in the modulation of the ACF of the random process. Taking the Fourier transformation on both sides of Eq. (55) :

$$P_{YY}(\Omega) = \frac{P_{XX}(\Omega + \Omega_c) + P_{XX}(\Omega - \Omega_c)}{4} \quad (56)$$

This result implies that modulation of a random process $X(t)$ with a random cosine results in a spectral upshift and a down-shift of the power spectrum $P_{XX}(\Omega)$.

Let us now look at the case where the random process $X(t)$ is modulated with a deterministic cosine. In specific let us look at the power spectrum of a *double sideband amplitude modulated* (DSB-AM) signal given by:

$$S_{DSBAM}(t) = m(t) \cos(\Omega_c t + \phi_o). \quad (57)$$

The message signal or baseband signal $m(t)$ (human-voice). Let us assume that the message signal $m(t)$ is a zero mean WSS process with ACF $R_{mm}(\tau)$. The mean of the random process $S_{DSBAM}(t)$ is therefore:

$$\mu_S(t) = E(m(t)) \cos(\Omega_c t + \phi_o) = 0. \quad (58)$$

The random process $S_{DSBAM}(t)$ is therefore first-order cyclostationary. The ACF of the modulated signal denoted $R_{SS}(t, t - \tau)$ is given by:

$$R_{SS}(t, t - \tau) = \frac{R_{mm}(\tau)}{2} [\cos(\Omega_c \tau) + \cos(2\Omega_c t - \Omega_c \tau + 2\phi_o)]. \quad (59)$$

Note that this ensemble ACF $R_{SS}(t, t - \tau)$ depends on both t and τ . This process is therefore not WSS. The ensemble ACF, however, is periodic in the variable t with fundamental period $T_o = \pi/\Omega_c$. The random process $S_{DSBAM}(t)$ is therefore second-order cyclostationary with parameter T_o . We can then evaluate the time-averaged ACF denoted $\tilde{R}_{SS}(\tau)$ via:

$$\tilde{R}_{SS}(\tau) = \frac{1}{T_o} \int_{-\frac{T_o}{2}}^{\frac{T_o}{2}} R_{SS}(t, t - \tau) dt = \frac{R_{mm}(\tau)}{2} \cos(\Omega_c \tau). \quad (60)$$

The corresponding power spectrum of DSB-AM can therefore be expressed as:

$$\tilde{P}_{SS}(\Omega) = \frac{P_{mm}(\Omega + \Omega_c) + P_{mm}(\Omega - \Omega_c)}{4}, \quad (61)$$

where $P_{mm}(\Omega)$ is the power spectrum of the baseband signal $m(t)$. As in the deterministic case, the baseband power spectrum has been up-shifted and down-shifted in frequency by Ω_c to give rise to the *upper sideband* (USB) and the *lower sideband* (LSB).

For a real message and WSS signal $m(t)$, the ACF is also real and even. It is easy to see that spectral components that are separated by $2\Omega_c = 2\pi/T_o$ are perfectly correlated (spectral redundancy). This will be of interest in co-channel communication systems where multiple users share the same carrier frequency. Note this result is analogous to the result that we obtained for the power spectrum of a random signal modulated with a random-cosine.