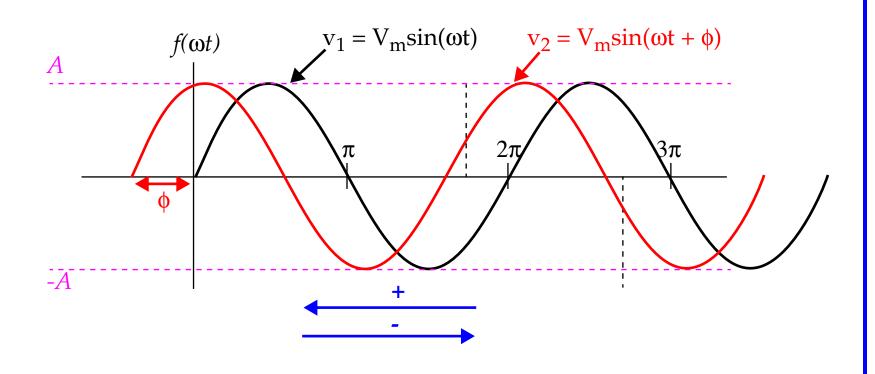
# **Properties of Sinusoids**

Relationship:

$$\omega = 2\pi f$$

$$f(t) = V_m \cos(\omega t + \phi)$$

 $\omega$  is the **angular frequency** in radians/sec  $\phi$  is the **argument** or **phase angle** (in radians) Period  $T = 2\pi/\omega = 1/f$ 





#### **Properties of Sinusoids**

Trig identities

$$\sin(\omega t + 90^{\circ}) = \cos(\omega t)$$

$$\sin(\omega t - 90^\circ) = -\cos(\omega t)$$

$$cos(\omega t + 90^{\circ}) = -sin(\omega t)$$

$$\cos(\omega t - 90^{\circ}) = \sin(\omega t)$$

$$\sin(\omega t + 180^{\circ}) = -\sin(\omega t)$$

$$cos(\omega t + 180^{\circ}) = -cos(\omega t)$$

$$\sin(A+B) = \sin A \cos B + \cos A \sin B$$

$$\sin(A - B) = \sin A \cos B - \cos A \sin B$$

$$\cos(A + B) = \cos A \cos B - \sin A \sin B$$

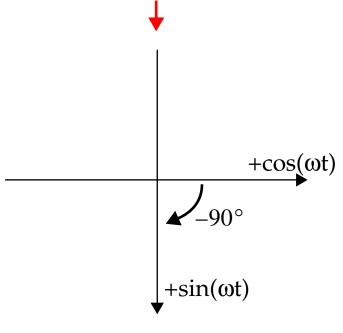
$$\cos(A - B) = \cos A \cos B + \sin A \sin B$$

Adding two sinusoids of same  $\omega$ 

$$A\cos(\omega t) + B\sin(\omega t) = C\cos(\omega t - \theta)$$

$$C = \sqrt{A^2 + B^2} \qquad \theta = \tan^{-1} \left(\frac{B}{A}\right)$$

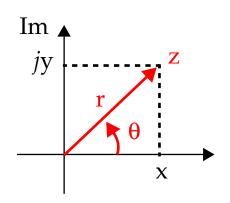
Don't confuse with complex plane described below



### **Complex Numbers**

Complex numbers: rectangular form

$$z = x + jy$$
$$j = \sqrt{-1}$$



$$\frac{1}{j} = -j$$

$$j^{2} = -1$$

$$j^{3} = -j$$

$$j^{4} = 1$$

Polar form

$$z = |z| \angle \theta \qquad \qquad z = x + jy = r \angle \theta = r\cos(\theta) + jr\sin(\theta)$$

$$r = \sqrt{x^2 + y^2} \qquad \qquad x = r\cos(\theta)$$

$$\theta = \tan^{-1}\left(\frac{y}{x}\right) \qquad \text{or} \qquad y = r\sin(\theta)$$

$$z = x + jy \implies \theta = \tan^{-1}\left(\frac{y}{x}\right)$$

$$z = -x + jy \implies \theta = 180^{\circ} - \tan^{-1}\left(\frac{y}{x}\right)$$

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# **Complex Numbers**

Complex conjugate

$$z^* = x - jy = r \angle -\theta = e^{-j\theta}$$

(Last form is exponential form)

Sum/difference: use rectangular form. Mult/div: use polar form

$$z_1 z_2 = r_1 r_2 \angle (\theta_1 + \theta_2)$$

$$\frac{z_1}{z_2} = \frac{r_1}{r_2} \angle (\theta_1 - \theta_2)$$

Exponential form

$$z = re^{j\theta}$$

(similar to polar form)

from Euler's formula

$$e^{j\theta} = \cos(\theta) + j\sin(\theta) \qquad \longrightarrow \qquad \cos(\theta) = Re(e^{j\theta})$$

$$e^{-j\theta} = \cos(\theta) - j\sin(\theta) \qquad \longrightarrow \qquad \sin(\theta) = Im(e^{j\theta})$$

Adding: 
$$\cos(\theta) = \frac{1}{2}(e^{j\theta} + e^{-j\theta})$$

Subtracting: 
$$\sin(\theta) = \frac{1}{2j} (e^{j\theta} - e^{-j\theta})$$



#### **Complex Numbers**

Useful identities

Assume 
$$z = x + jy = r \angle \theta$$
  
 $zz^* = x^2 + y^2 = r^2$   
 $\sqrt{z} = \sqrt{x + jy} = \sqrt{r}e^{j\theta/2} = \sqrt{r}\angle(\theta/2)$   
 $z^n = (x + jy)^n = r^n \angle n\theta = r^n e^{jn\theta} = r^n(\cos(n\theta) + j\sin(n\theta))$   
 $\ln(re^{j\theta}) = \ln(r) + \ln e^{j\theta} = \ln(r) + j\theta$   
 $e^{j\pi} = -1$   
 $e^{j2\pi} = e^{j0} = 1$   
 $e^{j\pi/2} = e^{j90^\circ} = j$  (IMPORTANT IDENTITIES FOR US)  
 $e^{j\pi/4} = e^{j45^\circ} = \sqrt{j} = \frac{(1+j)}{\sqrt{2}}$   $\sqrt{2j} = (1+j)$   
 $e^{-j\pi/2} = e^{j270^\circ} = -j$   $\sqrt{\frac{1}{j}} = \sqrt{-j} = \frac{(1-j)}{\sqrt{2}}$   
 $Re(e^{(\alpha + j\omega)t}) = Re(e^{\alpha t}e^{jwt}) = e^{\alpha t}\cos(\omega t)$   
 $Im(e^{(\alpha + j\omega)t}) = Im(e^{\alpha t}e^{jwt}) = e^{\alpha t}\sin(\omega t)$ 



Sines and cosines can be expressed in terms of *phasors*.

A phasor is a complex number that represents the amplitude and phase of a sinusoid.

Phasors based on Euler's identity

$$e^{j\theta} = \cos(\theta) + j\sin(\theta)$$
 and  $e^{-j\theta} = \cos(\theta) - j\sin(\theta)$ 

So  $cos(\theta)$  and  $sin(\theta)$  represent the Re and Im parts of  $e^{j\theta}$ .

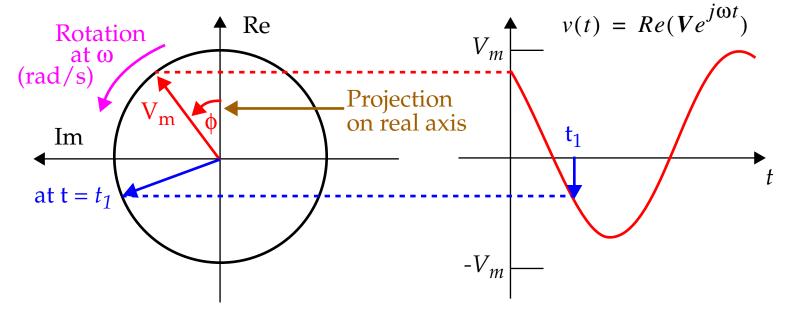
We can write a sinusoid

$$v(t) = V_m \cos(\omega t + \phi) = Re(V_m e^{j(\omega t + \phi)})$$
  
 $v(t) = Re(V_m e^{j\phi} e^{j\omega t})$  Arg includes phase and time part

Defining the phasor as the magnitude and the non-time part of the arg

$$v(t) = Re(Ve^{j\omega t}) \qquad V = V_m e^{j\phi} = V_m \angle \phi$$

The plot a rotating phasor:  $sinor Ve^{j\omega t}$ 



Note value of sinor at t = 0 is phasor **V** of the sinusoid v(t).

NOTE: the term  $e^{j\omega t}$  is implicitly present whenever a sinusoid is expressed as a phasor.

This is where the third parameter,  $\omega$ , is realized -- in the time part.

It is the third constant associated with the sinusoid.

To obtain the **sinusoid** associated with a **phasor**, multiply phasor **V** by  $e^{j\omega t}$  and take the Re part.

$$v(t) = Re(Ve^{j\omega t}) = Re(V_m e^{j(\omega t + \phi)}) = Re(V_m \cos(\omega t + \phi) + jV_m \sin(\omega t + \phi))$$
$$v(t) = V_m \cos(\omega t + \phi)$$

To obtain the **phasor** associated with a **sinusoid**, first write the sinusoid in cosine form (so it can be written as the real part of a complex number), and take out the time factor.

$$v(t) = V_m \cos(\omega t + \phi) = Re(V_m e^{j(\omega t + \phi)}) = Re(V_m e^{j\phi} e^{j\omega t}) = V_m e^{j\phi} = V_m \angle \phi$$

Phasors are complex, and therefore can be represented in rectangular, polar or exponential form.

They are **vectors** (magnitude and direction) and can be plotted as such.

The phasor domain is also known as the frequency domain.

Taking the derivative of a sinusoid

$$\frac{dv}{dt}(V_m\cos(\omega t + \phi)) = -\omega V_m\sin(\omega t + \phi) = \omega V_m\cos(\omega t + \phi + 90)$$

$$\frac{dv}{dt}(V_m\cos(\omega t + \phi)) = Re(\omega V_m e^{j\omega t} e^{j\phi} e^{j90^{\circ}}) = Re(j\omega V e^{j\omega t})$$

The relationships of derivative and integral

$$\frac{dv}{dv} \Leftrightarrow j\omega V$$

$$\int v dt \Leftrightarrow \frac{V}{i\omega}$$

These relationships are useful in computing the steady-state response (where we don't need to know the initial values).

Phasors are also useful for summing sinusoids of the same frequency.

Important FACTs to remember:

- v(t) is time dependent, while **V** is NOT.
- v(t) is ALWAYS real, while **V** is generally complex.
- Phasor analysis applies only when frequency is **constant** and allows two or more sinusoids to be manipulated ONLY if they have the same frequency.

## Examples

$$v = -4\sin(30t + 50^{\circ})V \longrightarrow -\sin A = \cos(A + 90^{\circ})$$

$$v = 4\cos(30t + 50^{\circ} + 90^{\circ}) = 4\cos(30t + 140^{\circ})$$

$$V = 4\angle 140^{\circ}V$$

$$I = -3 + j4A \qquad \text{(Phasors can be in any of the three forms)}$$

$$I = 5\angle 126.87^{\circ}$$

$$i(t) = 5\cos(\omega t + 126.87^{\circ})A$$

$$V = j8e^{-j20^{\circ}}V \qquad \text{since } j = 1\angle 90^{\circ}$$

$$V = j8\angle -20^{\circ} = (1\angle 90^{\circ})(8\angle -20^{\circ}) = 8\angle 90^{\circ} - 20^{\circ} = 8\angle 70^{\circ}$$

$$v(t) = 8\cos(\omega t + 70^{\circ})V$$

Used to solve an integrodifferential equation

Solve for i(t)

$$4i + 8 \int i dt - 3 \frac{di}{dt} = 50 \cos(2t + 75^{\circ})$$

$$4I + \frac{8I}{j\omega} - 3j\omega I = 50 \angle 75^{\circ}$$
Convert to phasor domain
$$I(4 - j4 - j6) = 50 \angle 75^{\circ}$$

$$\omega \text{ is 2 and- } 1/j = -j.$$

$$I = \frac{50 \angle 75^{\circ}}{4 - j10} = \frac{50 \angle 75^{\circ}}{10.77 \angle (-68.2)^{\circ}} = 4.642 \angle 143.2^{\circ}$$

$$m = \sqrt{4^2 + 10^2} = 10.77$$

$$\phi = \tan^{-1} \left(\frac{10}{4}\right) = 360^\circ - 68.2^\circ = 291.8^\circ$$

$$\phi = 75^\circ - 291.8^\circ = -216.8^\circ + 360 = 143.2^\circ$$

So far we've shown how to represent a voltage and current in phasors.

What about the passive elements, R, L and C

## Phasors Applied to R, L and C circuits

Need to transform the current/voltage relationships for R, L and C into the frequency domain.

#### For **R**

$$i = I_m \cos(\omega t + \phi)$$

$$v = iR = RI_m \cos(\omega t + \phi)$$

$$V = RI_m \angle \phi$$

$$I = I_m \angle \phi$$

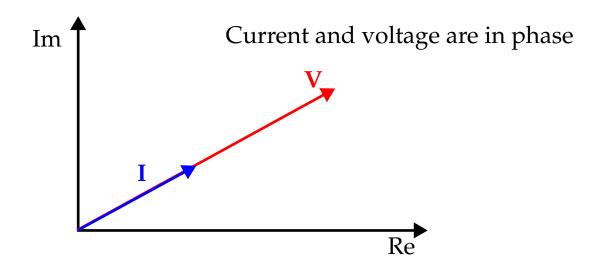
$$V = RI$$

Assume current through a resistance **R** 

And voltage across it

The phasor for this voltage

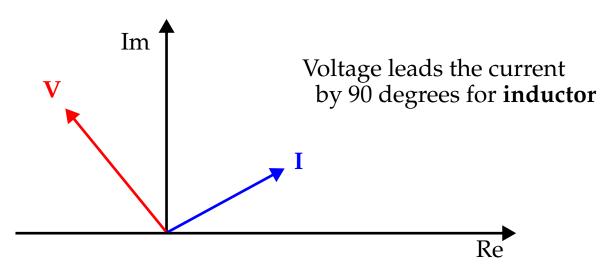
Phasor for the current



#### Phasors Applied to R, L and C circuits

For L

$$i = I_m \cos(\omega t + \phi)$$
 Assume current through inductor L  
 $v = L \frac{di}{dt} = -\omega L I_m \sin(\omega t + \phi)$  Voltage across it  
 $v = L \frac{di}{dt} = \omega L I_m \cos(\omega t + \phi + 90^\circ)$   $-\sin(A) = \cos(A + 90^\circ)$   
 $V = \omega L I_m e^{j(\phi + 90^\circ)} = \omega L I_m e^{j\phi} e^{j90^\circ}$  Into a phasor  
 $V = j\omega L I$   $e^{j90^\circ} = j$ 



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The j term translates into a phase shift in the time domain.

### **Impedance**

For **C** 

$$i = C \frac{dv}{dt}$$
  $\longrightarrow$   $V = \frac{I}{j\omega C}$  or  $I = j\omega CV$ 

Here, current leads voltage by 90 degrees

The first form relates *V* and *I* similar to resistance.

These can be written in terms of their phasor voltage to phasor current

$$\frac{V}{I} = R$$

$$\frac{V}{I} = j\omega L \qquad \longrightarrow \qquad Z = \frac{V}{I} \qquad \text{Impedance } Z \text{ is the ratio of phasor voltage to current in } \Omega S$$

$$\frac{V}{I} = \frac{1}{j\omega C} \qquad \text{or} \qquad \frac{V}{I} = \frac{-j}{\omega C}$$

**NOTE:** the impedance **Z** is NOT a phasor -- it does not correspond to a sinusoidally varying quanity.

## **Impedance**

Example

$$v = 12\cos(60t + 45^{\circ})$$
  
 $L = 0.1 \text{ H}$  Find steady-state current  
 $V = j\omega LI$  with  $w = 60 \text{ rad/s}$  and  $V = 12\angle 45^{\circ}$   
 $I = \frac{V}{i\omega L} = \frac{12\angle 45^{\circ}}{i60(0.1)} = \frac{12\angle 45^{\circ}}{6\angle 90^{\circ}} = 2\angle -45^{\circ}A$ 

So we see the magnitude of the impedance is 6  $\Omega$ s and the inductor causes a phase shift in the current (it lags by 90 degrees).

$$Z = R + jX$$
  $R = Re(Z)$  resistance  
 $Z = R - jX$   $X = Im(Z)$  reactance

Reactance can be *negative* (capacitive, current lags) or *positive* (inductive).

Other forms

$$Z = R + jX = |Z| \angle \theta$$
  $|Z| = \sqrt{R^2 + X^2}$   $\theta = \tan^{-1} \left(\frac{X}{R}\right)$   $R = |Z|\cos(\theta)$   $X = |Z|\sin(\theta)$ 

#### **Admittance**

Admittance Y is the reciprocal of impedance, measured in siemens (S)

$$Y = \frac{1}{Z} = \frac{I}{V}$$

$$Y = G + jB$$

G = Re(Y) is called conductance (mohms)

B = Im(Y) is called susceptance

$$G + jB = \frac{1}{R + jX}$$

$$G + jB = \frac{1}{R + jX} \frac{R - jX}{R - jX} = \frac{R - jX}{R^2 + X^2}$$
 Relating impedance with admittance

$$G = \frac{R}{R^2 + X^2} \qquad B = \frac{-X}{R^2 + X^2}$$

$$B = \frac{-X}{R^2 + X^2}$$

So, G does NOT equal 1/R as was true for resistive circuits.

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### **Examples**

$$V_s = 10(\angle 0^\circ)V$$

$$v_s = 10\cos(4t) + 0.1F$$

The impedance is

$$Z = R + \frac{1}{j\omega C} = 5 + \frac{1}{j4(0.1)} = 5 - j2.5 \Omega$$

And current

$$I = \frac{V_s}{Z} = \frac{10\angle 0^{\circ}}{5 - j2.5} = \frac{10(5 + j2.5)}{5^2 + 2.5^2} = 1.6 + j0.8 = 1.789\angle 26.57^{\circ}$$
OR

$$I = \frac{V_s}{Z} = \frac{10\angle 0^{\circ}}{5.6\angle -26.57} = 1.789\angle 26.57^{\circ}$$

Voltage across capacitor

$$V = IZ_C = \frac{I}{j\omega C} = \frac{1.789\angle 26.57^{\circ}}{j4(0.1)} = \frac{1.789\angle 26.57^{\circ}}{0.4\angle 90^{\circ}} = 4.47\angle -63.43^{\circ}$$

Time domain

$$i(t) = 1.789\cos(4t + 26.57^{\circ})$$

$$v(t) = 4.47\cos(4t - 63.43^{\circ})$$